

General Description

The MAX3801 is a +3.3V adaptive cable equalizer designed for coaxial and twin-axial cable point-to-point communications applications. The equalizer includes differential CML data inputs and outputs, a loss-of-signal (LOS) output, and a cable integrity monitor (CIM) output.

The adaptive cable equalizer is capable of equalizing differential or single-ended signals at data rates up to 3.2Gbps. It automatically adjusts to attenuation caused by skin-effect losses of up to 30dB at 1.6GHz. The equalizer effectively extends the usable length of copper cable in high-frequency interconnect applications.

The MAX3801 is available in a 24-pin QFN package with exposed pad and consumes only 125mW at +3.3V.

Applications

High-Speed Links in Communications and Data Systems

Backplane and Interconnect Applications SDH/SONET Transmission Equipment

Features

- ♦ Single +3.3V Operation
- ◆ Typical Power Dissipation = 125mW at +3.3V
- ♦ Data Rates Up to 3.2Gbps
- **♦** Equalizer Automatically Adjusts for Different **Cable Lengths**
- ♦ 0dB to 30dB Equalization at 1.6GHz (3.2Gbps)
- ♦ Loss-of-Signal (LOS) Indicator
- **♦ Cable Integrity Monitor (CIM)**
- ♦ On-Chip Input and Output Terminations
- **♦ Low External Component Count**
- ♦ Operating Temperature Range = 0°C to +85°C
- **♦** ESD Protection on Inputs and Outputs

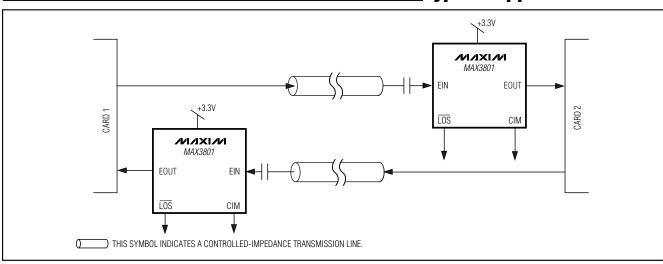
Ordering Information

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	PART	TEMP. RANGE	PIN-PACKAGE
ſ	MAX3801UGG	0°C to +85°C	24 QFN-EP*

^{*}EP = exposed pad

Pin Configuration appears at end of data sheet.

Typical Application Circuit



MIXIM

Maxim Integrated Products 1

ABSOLUTE MAXIMUM RATINGS

Supply Voltage, V _{CC} 0.5V to +6.0V	Operating Ambient Temperature Range0°C to +85°C
Voltage at \overline{LOS} , CIM+, CIM0.5V to (V _{CC} + 0.5V)	Storage Temperature Range55°C to +150°C
Voltage at EIN+, EIN(V _{CC} - 1V) to (V _{CC} + 0.5V)	Lead Temperature (soldering, 10s)+300°C
Current Out of EOUT+, EOUT25mA	
Continuous Power Dissipation (T _A = +85°C)	
24-Pin OFN-FP (derate 25.1mW/°C above +85°C) 1630mW	

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

DC ELECTRICAL CHARACTERISTICS

 $(V_{CC} = +3.14V \text{ to } +3.46V, T_A = 0^{\circ}\text{C} \text{ to } +85^{\circ}\text{C}.$ Typical values are at $V_{CC} = +3.3V \text{ and } T_A = +25^{\circ}\text{C}, \text{ unless otherwise noted.})$

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
Supply Current	Icc	Includes external load current		37	60	mA
INPUT SPECIFICATIONS						
Minimum Cable Input (Differential)		3.2Gbps, 30dB cable loss at 1.6GHz (Note 1)		650	700	mVp-p
Maximum Cable Input (Differential)				1100		mVp-p
Input Impedance		Single-ended	40	53	62.5	Ω
OUTPUT SPECIFICATIONS						
Output Voltage (Differential)		(Note 2)	500		1000	mVp-p
Output Impedance		Single-ended	50	65	75	Ω
Voltage at CIM Output (Differential)	V _{CIM}	No external load, V _{CIM} = (V _{CIM+}) - (V _{CIM-})	-0.5		+0.5	Vp-p
Voltage at CIM Output (Single-Ended)	V _{CIM+} , V _{CIM-}	No external load	0.5		V _C C - 0.5	V
V-ll		Output high (Note 3)	2.4			
Voltage at LOS		Output low (Note 3)			0.4	V
Output Common-Mode Voltage		Each output DC-coupled 50Ω to V_{CC}		V _{CC} - 0.2		V

AC ELECTRICAL CHARACTERISTICS

 $(V_{CC} = +3.14 \text{V to } +3.46 \text{V}, T_A = 0^{\circ}\text{C} \text{ to } +85^{\circ}\text{C}. \text{ Typical values are at } V_{CC} = +3.3 \text{V and } T_A = +25^{\circ}\text{C}, \text{ unless otherwise noted.})$ (Note 4)

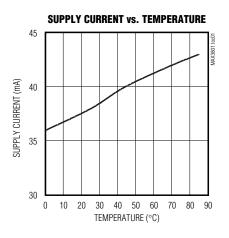
PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
Maximum Input Data Rate			3.2			Gbps
		0dB cable loss (Note 6)		120	240	
Residual Jitter (Note 5)		24dB cable loss (Note 6)		140	240	mUlp-p
		30dB cable loss (Note 6)		100	200	
Output Edge Speed		20% to 80%		64	90	ps
Input Return Loss (Single-Ended)		≤3.2GHz		15		dB
Output Return Loss (Single-Ended)		≤3.2GHz		15		dB
Equalization Compensation		1.6GHz (skin-effect losses only)	30			dB
Equalization Time Constant				5		μs

- **Note 1:** Minimum cable input for \overline{LOS} to assert high.
- **Note 2:** Input voltage within specification limits, 50Ω to V_{CC} at each output.
- **Note 3:** $100k\Omega$ load to ground.
- Note 4: AC electrical characteristics are guaranteed by design and characterization.
- Note 5: Includes random jitter and deterministic jitter.
- **Note 6:** Differential cable input voltage = 700mVp-p, 3.2Gbps 2¹³-1 PRBS with 100 consecutive ones and 100 consecutive zeros substituted. Cable loss is due to skin effect only.

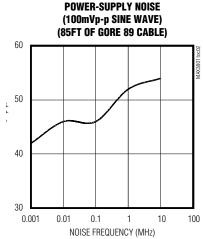
Typical Operating Characteristics

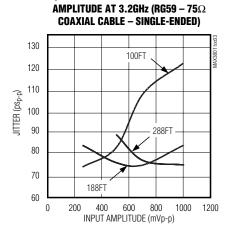
 $(V_{CC} = +3.3V, \text{ all jitter measurements done at } 3.2\text{Gbps}, 700\text{mV} \text{ cable input with } 2^{13}\text{-}1 \text{ PRBS pattern with } 100 \text{ consecutive ones and } 100 \text{ consecutive zeros substituted, } T_A = +25^{\circ}\text{C}.$ **Note:** Test pattern produces near worst-case jitter results. Results will vary with pattern, unless otherwise noted.)

EOUALIZER RESIDUAL JITTER vs.

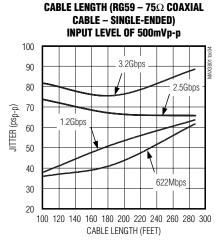


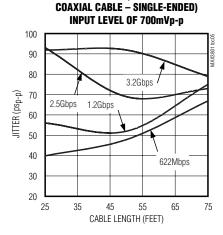
EQUALIZER RESIDUAL JITTER vs.





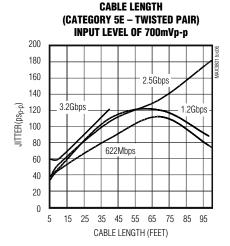
EQUALIZER RESIDUAL JITTER vs. INPUT





EQUALIZER RESIDUAL JITTER vs.

CABLE LENGTH (RG179B - 75 Ω

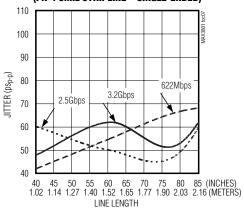


EQUALIZER RESIDUAL JITTER vs.

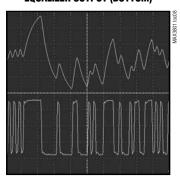
Typical Operating Characteristics (continued)

($V_{CC} = +3.3V$, all jitter measurements done at 3.2Gbps, 700mV cable input with 2^{13} -1 PRBS pattern with 100 consecutive ones and 100 consecutive zeros substituted, $T_A = +25$ °C. **Note:** Test pattern produces near worst-case jitter results. Results will vary with pattern, unless otherwise noted.)

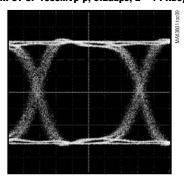
EQUALIZER RESIDUAL JITTER vs. LINE LENGTH (FR-4 6MIL STRIPLINE – SINGLE-ENDED)



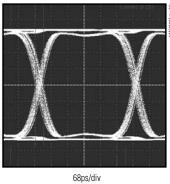
EQUALIZER INPUT AFTER 115FT OF CABLE (TOP) EQUALIZER OUTPUT (BOTTOM)



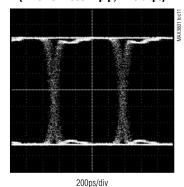
EQUALIZER OUTPUT EYE DIAGRAM AFTER 288 FT OF RG59 CABLE (INPUT OF 1000mVp-p, 3.2Gbps, 2²³-1 PRBS)



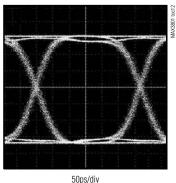
EQUALIZER OUTPUT EYE DIAGRAM AFTER 100FT OF 75Ω RG179 CABLE (2.5Gbps, SINGLE-ENDED, 2^7 -1 PRBS)



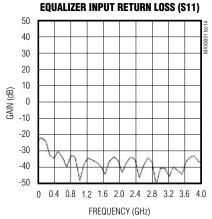
EQUALIZER OUTPUT EYE DIAGRAM AFTER 70FT OF CATEGORY 5E CABLE (INPUT OF 1000mVp-p, 1.25 Gbps)



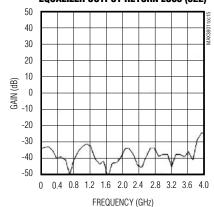
EQUALIZER OUTPUT EYE DIAGRAM AFTER 115FT OF 50Ω GORE 89 CABLE (INPUT OF 1000mVp-p, 3.2Gbps)



200μ3/μ1



EQUALIZER OUTPUT RETURN LOSS (\$22)



MIXIM

Pin Description

PIN	NAME	FUNCTION
1, 3, 7, 12, 16, 18, 19, 24	GND	Ground
2, 4, 8, 11, 17, 20, 23	Vcc	Power Supply
5	CIM-	Negative Cable Integrity Monitor (CIM) Output
6	CIM+	Positive Cable Integrity Monitor (CIM) Output
9	EOUT-	Negative Equalizer Output, CML
10	EOUT+	Postive Equalizer Output, CML
13, 14	N.C.	No connection. Leave unconnected.
15	LOS	Equalizer Loss-of-Signal Output, Active-Low
21	EIN+	Postive Equalizer Input, CML
22	EIN-	Negative Equalizer Input, CML
EP	Exposed Pad	Ground. The exposed pad must be soldered to the circuit board ground for proper thermal and electrical operation.
СР	Corner Pins	Ground. The corner pins must be soldered to supply ground.

Detailed Description

The adaptive cable equalizer accepts differential CML input data at rates up to 3.2Gbps and is capable of equalizing differential or single-ended signals. It automatically adjusts to attenuation levels of up to 30dB at 1.6GHz (due to skin-effect losses in copper cable). The equalizer consists of a CML input buffer, a loss-of-signal detector, a flat response amplifier, a skin-effect compensation amplifier, a current-steering network, a dual power-detector feedback loop, an output limiting amplifier, and a CML output buffer (Figure 1).

General Theory of Operation

The shape of the power spectrum of a random bit stream can be described by the square of the sinc function, where sinc $f=(\sin\pi f)\,/\,\pi f.$ For sufficiently long bit patterns (nonrandom bit streams), $\text{sinc}^2(f)$ is a good approximation. From the shape of the $\text{sinc}^2(f)$ function, we can estimate the ratio of the power densities at any two frequencies. The MAX3801 adaptive equalizer employs this principle by incorporating a feedback loop that continuously monitors the power at two frequencies and dynamically adjusts the equalizer to maintain the correct power ratio.

CML Input and Output Buffers

The input and output buffers are implemented using current-mode logic (CML). Equivalent circuits are shown in Figures 2 and 3. For details on interfacing with CML, see Maxim application note HFAN-1.0, *Interfacing Between CML, PECL, and LVDS.*

Flat Response and Skin-Effect Compensation Amplifiers

The buffered input waveform is fed equally to two amplifiers—the flat response amplifier and the skineffect compensation amplifier. The flat response amplifier has a constant gain over the entire frequency range of the device, and the skin-effect compensation amplifier has a gain characteristic that approximates the inverse of the skin-effect attenuation inherent in copper cable. The skin-effect attenuation, in dB per unit length, is proportional to the square root of the frequency. The output currents from the two amplifiers are supplied to the current-steering network. Note that when $\overline{\text{LOS}}$ asserts low, equalization is minimized.

Current-Steering Network

The function of the current-steering network is to combine adjustable quantities of the output currents from the flat response and skin-effect compensation ampli-

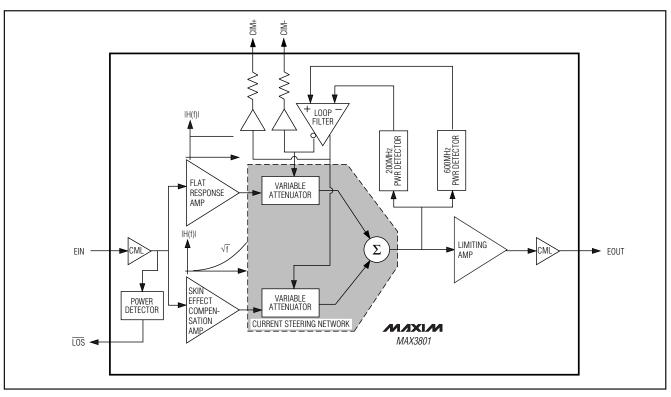


Figure 1. Functional Diagram

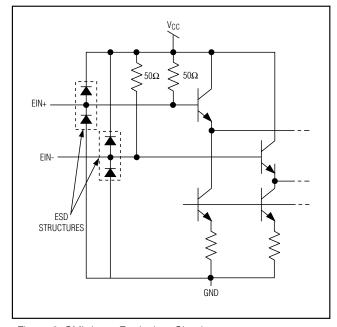


Figure 2. CML Input Equivalent Circuit

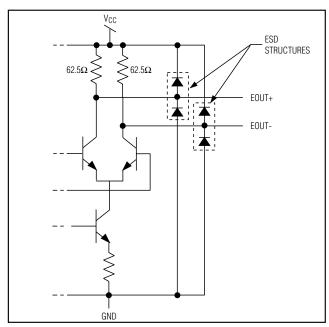


Figure 3. CML Output Equivalent Circuit

fiers to achieve a desired current ratio. The ratio adjustment is controlled by the dual power-detector feedback loop.

The current-steering network is implemented with a pair of variable attenuators that feed into a current-summing node. The variable attenuators are used to attenuate the output currents of the flat response and skin-effect compensation amplifiers under control of the dual power-detector feedback loop. The outputs of the two attenuators are combined at the summing node and then fed to the output limiting amplifier and the feedback loop.

Dual Power-Detector Feedback Loop

The output of the current-steering network is applied to the inputs of two frequency-specific power detectors. One of the power detectors is tuned to 200MHz and the other is tuned to 600MHz. The outputs of the two power detectors are applied to the inverting (200MHz power detector) and noninverting (600MHz power detector) inputs of the differential loop amplifier. The differential outputs of the loop amplifier control the variable attenuators in the current-steering network.

Output Limiting Amplifier

The output limiting amplifier amplifies the signal from the current-steering network to achieve the specified output voltage swing.

Applications Information

Refer to Maxim application note HFDN-10.0, *Equalizing Gigabit Copper Cable Links with the MAX3800* (available at *www.maxim-ic.com*) for additional applications information.

Cable Integrity Monitor (CIM)

The differential CIM output current is directly proportional to the output current of the loop amplifier (which controls the current-steering network—see *Detailed Description*). This is an analog current output that indicates the amount of equalization that is being applied. A convenient way to monitor the CIM current is to connect a $100\text{k}\Omega$ resistor from each of the CIM outputs to ground, and then measure the voltage at the CIM pins.

The amount of equalization (and thus the CIM output level) is affected by various factors, including cable type, cable length, signal bandwidth, etc.

Loss-of-Signal (LOS) Output

Loss-of-signal is indicated by the $\overline{\text{LOS}}$ output. A low level on $\overline{\text{LOS}}$ indicates that the equalizer input signal

power has dropped below a threshold. The $\overline{\text{LOS}}$ output indicates a loss of signal. When the equalizer no longer detects a signal from the channel, the $\overline{\text{LOS}}$ output goes low. When there is sufficient input voltage to the channel (typically greater than 650mV), $\overline{\text{LOS}}$ is high. The $\overline{\text{LOS}}$ output is suitable for indicating problems with the transmission link caused by, for example, a broken cable, a defective driver, or a lost connection to the equalizer.

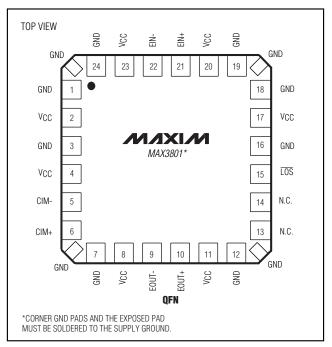
Single-Ended Operation

For single-ended operation of the equalizer, connect the unused input to ground through a series combination of a capacitor (of equal value to other AC-coupling capacitors) and a 50Ω resistor. Note that the MAX3801 is specified for differential operation. The effective range of equalization for single-ended use is approximately 4dB to 30dB at 1.6GHz.

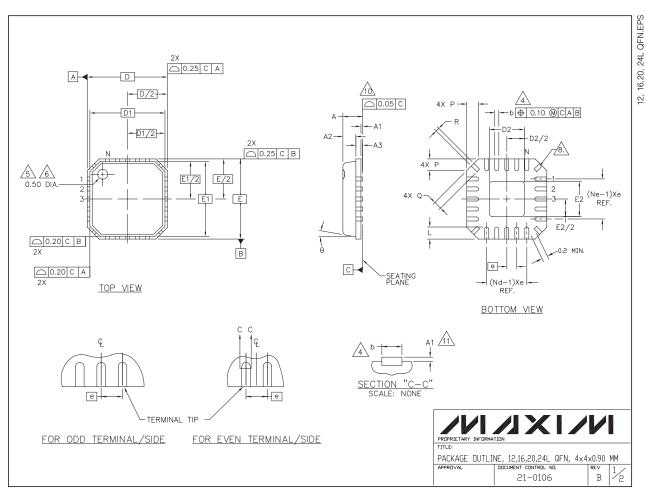
Layout Considerations

The MAX3801's performance can be significantly affected by circuit-board layout and design. Use good high-frequency design techniques, including minimizing ground inductance and using fixed-impedance transmission lines for the high-frequency data signals. Power-supply decoupling capacitors should be placed as close as possible to $V_{\rm CC}$.

Pin Configuration



Package Information



MIXIM

Package Information (continued)

0.20 REF 4.00 BS0 3.75 BS0

0.42

0.17

0.60

MIN.

0.13

NOTES:

- 1. DIE THICKNESS ALLOWABLE IS 0.305mm MAXIMUM (.012 INCHES MAXIMUM)
- 2. DIMENSIONING & TOLERANCES CONFORM MUST TO ASME Y14.5M. 1994.

Ad is the number of terminals.

No is the number of terminals in X-direction & No is the number of terminals in Y-direction.

4. DIMENSION & APPLIES TO PLATED TERMINAL AND IS MEASURED

BETWEEN 0.20 AND 0.25mm FROM TERMINAL TIP.

THE PIN #1 IDENTIFIER MUST BE EXISTED ON THE TOP SURFACE OF THE PACKAGE BY USING INDENTATION MARK OR OTHER FEATURE OF PACKAGE BODY.

 $\sqrt{6\lambda}$ exact shape and size of this feature is optional.

7. ALL DIMENSIONS ARE IN MILLIMETERS.

8. THE SHAPE SHOWN ON FOUR CORNERS ARE NOT ACTUAL I/O.

9. PACKAGE WARPAGE MAX 0.05mm.

APPLIED FOR EXPOSED PAD AND TERMINALS.

EXCLUDE EMBEDDING PART OF EXPOSED PAD FROM MEASURING.

11 APPLIED ONLY FOR TERMINALS.

12	12. MEETS JEDEC MO220.																		
S M B	PITCH MIN.	VARIAT	ION A	N _O T _F	S M B	PITCH MIN.	VARIAT	ION B	N _O T _F	S M B O	PITCH MIN.	VARIAT	TON C	N _O T _F	SYMBO.	PITCH MIN.	VARIAT	ION D MAX.	No _T
e		0.80 BSC			e		0.65 BSC	1417 17 1		e		0.50 BSC			e		0.50 BSC	1717 17 1.	
N		12		3	N		16		3	N		20		3	N		24		3
Nd		3		3	Nd		4		3	Nd		5		3	Nd		6		3
Ne		3		3	Ne		4		3	Ne		5		3	Ne		6		3
L	0.50	0.60	0.75		L	0.50	0.60	0.75		L	0.50	0.60	0.75		L	0.30	0.40	0.55	
Ь	0.28	0.33	0.40	4	Ь	0.23	0.28	0.35	4	Ь	0.18	0.23	0.30	4	b	0.18	0.23	0.30	4
Q	0.30	0.40	0.65		Q	0.30	0.40	0.65		Q	0.30	0.40	0.65		Q	0.00	0.20	0.45	

SYMBOLS			D2			E2	NOTE	
		MIN	NOM	MAX	MIN	NOM	MAX	
EXPOSED PAD	Α	1.95	2.10	2.25	1.95	2.10	2.25	
VARIATIONS	В	1.55	1.70	1.85	1.55	1.70	1.85	

EXAMPLE: WE CAN CALL VARIATION "BB" FOR 16 TERMINAL QFN WITH 1.70x1.70 mm NOMINAL EXPOSED PAD DIMENSION. THE FORMER ONE IN VARIATION IS FOR PITCH VARIATION AND THE LATTER ONE IS FOR EXPOSED PAD VARIATION.

PROPRIETARY INFORMA				/
TITLE:				
PACKAGE DUTLI	NE, 12,16,20,24L	QFN, 4×4	×0.90	MM
APPROVAL	DOCUMENT CONTROL N	Ю.	REV	2/
	21-0106		l R	1/2

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